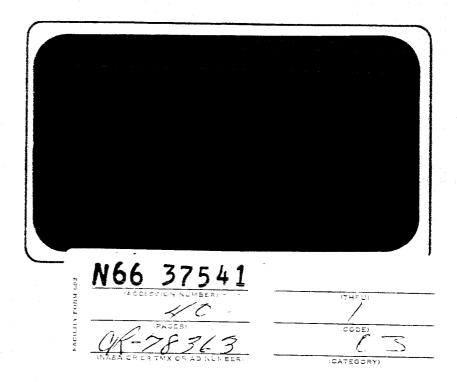


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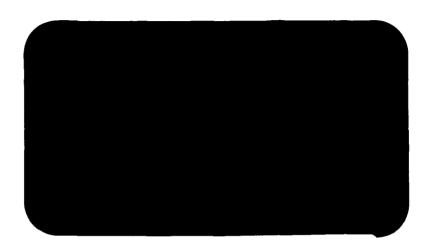


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Nickel-Cadmium Battery Reconditioner

Assignment 61 501 MEL R&D Report 183/66 August 1966

> Ву Floyd E. Ford

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#### ABSTRACT

A device has been designed for automatically reconditioning a nickel-cadmium battery. Individual cell voltage detectors and/or a battery terminal voltage are utilized to determine when reconditioning is necessary. Reconditioning is accomplished by discharging each cell to nearly 0 volts, then recharging the battery at a small current. Controlled discharge of the battery is accomplished by individual cell shunting transistors and a battery current shunt transistor. Cell reversal is prevented by cutting off the battery dump current when a cell voltage reaches a low limit.

author

#### REFERENCE

(a) Potter, Nelson and Richard Morrison, "Two Level Voltage Limiter," Goddard Space Flight Center Rept X-716-66-7, Jan 1966

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#### NICKEL-CADMIUM BATTERY RECONDITIONER

#### 1.0 INTRODUCTION

The U. S. Navy Marine Engineering Laboratory was requested by NASA (Goddard Space Flight Center (GSFC)) as part of Contract S-12730-G. Amendment 5 of 20 April 1965, to design and develop a device operating within a battery test system for automatically reconditioning nickel-cadmium cells. This report presents a description of the device designed to perform this task.

1.1 Development Task. When a nickel-cadmium battery is subjected to extensive cycling, degradation in the ampere-hour capacity of each cell occurs. Degradation takes place in the form of a "memory effect; i.e. after repetitive cycling of a nickel-cadmium to a given depth of discharge, the cell "remembers" this depth and will deliver only that amount of ampere-hour capacity, instead of its rated capacity. number of cycles to which a battery may be subjected before the loss in ampere-hour capacity is critical varies among batteries of the same capacity, and with environment and depth of discharge. When the memory effect has taken place, it is indicated by a low battery voltage and/or a low cell voltage near the end of the discharge cycle. Previous tests on nickel-cadmium cells have indicated that the ampere-hour capacity of a battery can be restored by "bleeding" each cell down to approximately 0 volts and then recharging the battery at a current that is very low (approximately C/10 current).

The "memory effect" on each cell varies from cell to cell and results in cell unbalance in a battery. Reconditioning removes this capacity unbalance by providing the cells with a common reference condition for recharging.

It is evident that reconditioning a battery pack of nickelcadmium cells results in an extended life for the battery and provides for greater utilization of a battery's ampere-hour capacity. This greater efficiency would permit the use of a smaller ampere-hour battery pack for a given satellite payload

<sup>1</sup> Referred to as reconditioning

<sup>2</sup>C = cell capacity in ampere-hours; 10 = number of hours for charging

capability. The ultimate goal in developing this nickel-cadmium battery reconditioner, originally designed to operate in a laboratory battery test system, is adaptation to satellite power systems.

1.2 Problem Background. The method presently used to recondition nickel-cadmium batteries is designed to count the number of cycles to which a battery has been subjected and then automatically to discharge the battery by placing a load resistor across each cell.

The disadvantage of this method is twofold. First, since the number of cycles, and not a low voltage condition, is used to determine when reconditioning is to take place, the battery will be reconditioned irrespective of the state of degradation of the cells. Secondly, to prevent reverse charging of a cell, relays switch individual load resistors across each cell to discharge the battery. This requires relay contacts in series with each cell of the battery. The new nickel-cadmium battery reconditioner is a voltage detection device that utilizes a low voltage condition on the battery terminals or on a cell to indicate when degradation of the battery is such that reconditioning is necessary. Since the number of cycles to which a battery can be subjected before severe degradation occurs varies widely with environment, the voltage detection method is superior to the method of counting cycles. The use of transistors as a shunt load for each cell eliminates the need for relays and associated contacts to switch load resistors across each cell.

#### 2.0 OPERATION

The following section provides a general description of the operation of the nickel-cadmium battery reconditioner and the battery test system it is used with. A detailed description of each circuit and its operation may be found in Section I of Appendix A. Section II of Appendix A is a detailed description of component. Section III contains the recommended adjustment and calibration procedure.

2.1 <u>Functional Description</u>. The nickel-cadmium battery reconditioner can best be described by the functional diagram shown in Figure 1. These functional units are: (1) battery under voltage detector and timer control, (2) cell voltage detector

and control circuit, (3) battery current dump transistor, (4) individual cell-shunt transistors, and (5) a cycle number control. The orbit time control, shunt regulator, charger, and spacecraft load are shown to illustrate the relationship of the battery reconditioner to the rest of the battery test system.

The charger, spacecraft load, and shunt regulator are connected in parallel. Normally, the power from the charger is supplied to the battery and load during the daylight period of orbit. In the dark period, the battery supplies the power demands of the load. If degradation of the battery has taken place (loss of ampere hour capacity), a low voltage condition on a cell or the battery will exist near the end of the discharge period. If the voltage of a cell or the battery falls below the respective preset limit, the battery is switched from the load to controlled discharge.

In controlled discharge, the battery-current dump transistor discharges the battery, and the cell-shunt transistors are biased "on" to discharge each cell. The current in the dump transistor is controlled by the voltage of the lowest cell of the battery. Cutoff is obtained when a cell reaches a preselected lower limit. This cutoff prevents reverse charging of that cell. The shunt transistors continue to deplete each cell of the remaining ampere-hour capacity, thereby assuring that the voltage of all cells will be reduced to near zero.

When the time control notes that sufficient time has elapsed for all cells to discharge, the timer removes the bias from the cell shunt transistors, opens the spacecraft load circuit, and places the battery in parallel with the charger and shunt regulator. The charger provides a constant current to recharge the battery to a voltage level determined by the shunt regulator, reference (a). Sufficient time for charging is preset into the timer control. When the battery is completely charged, the load is returned to the battery and charger. Normal cycling of the battery is resumed by the orbit time control with the battery reconditioner remaining in standby status. The cycle number control counts the number of cycles completed on the battery when a low voltage condition is reached. If desired, this device may also be used to switch the battery to controlled discharge after a preselected number of cycles has been completed.

2.2 Discussion. Since the initial purpose of the design of the nickel-cadmium battery reconditioner was to determine its feasibility through use in a battery test system, a number of parameters and operating conditions were made selective so that optimum conditions for system operation could be determined. These include the variable parameter adjustments which are: (1) battery low-voltage trip level (3 to 6 volts), (2) minimum cell voltage for zero-battery dump current (25 to 500 millivolts), (3) transistor-dump current (0.1 to 0.5 amperes), (4) cell shunt transistor current (50 to 750 milliamperes at 1 volt), and (5) sensitivity of dump transistor to control circuit. The following selective methods of switching to controlled discharge are provided: (1) low cell voltage, (2) low battery voltage, (3) cycle number limit, (4) low cell voltage or cycle-number limit, and (5) either low cell voltage, low battery voltage, or cycle-number limit. A means for bypassing the dump transistor, and using only the cell shunt transistors to discharge the battery, is also provided.

The added versatility obtained from these flexible characteristics allows this device to be used to obtain its own optimum operating conditions, as well as those for the battery pack. Although this instrument is designed for a 5-cell battery, the concept can be expanded to operate with any number of cells.

#### 3.0 RESULTS

The characteristics obtained from a typical battery reconditioning operation are illustrated in Figure 2. (1) A 5-cell, 6 ampere-hour nickel cadmium battery was being cycled in a battery test system. The cell voltage of four of the five cells was recorded. A low cell voltage on cell No. 1 is noted during discharge of the battery. At time T = 15 minutes on the graph, the reconditioner switched the battery to the dump transistor ( $Q_6$ ) of Figure 1. The battery current through the dump transistor is reduced to approximately 0 as the voltage of cell No. 1 is reduced to a preset low limit. The remaining cells have their residual ampere-hour capacity removed by the individual cell shunt transistors ( $Q_6$  of Figure 1). The area under the curves for each cell provides a graphic illustration of cell unbalance within the battery.

Data obtained from GSFC, Greenbelt, Maryland

#### 4.0 CONCLUSION

Two prototypes of the nickel-cadmium battery reconditioner have been built and delivered to NASA, GSFC, Greenbelt, Maryland. The first prototype was used to determine the feasibility of such a device, while the second one was built to serve as a model for production of the device. Photographs of the second unit are shown in Figures 3, 4, and 5.

The first prototype has been operating satisfactorily in a battery test system since December 1965. Results indicate that the voltage method of determining when a battery should be reconditioned is far superior to the number-of-cycles method.

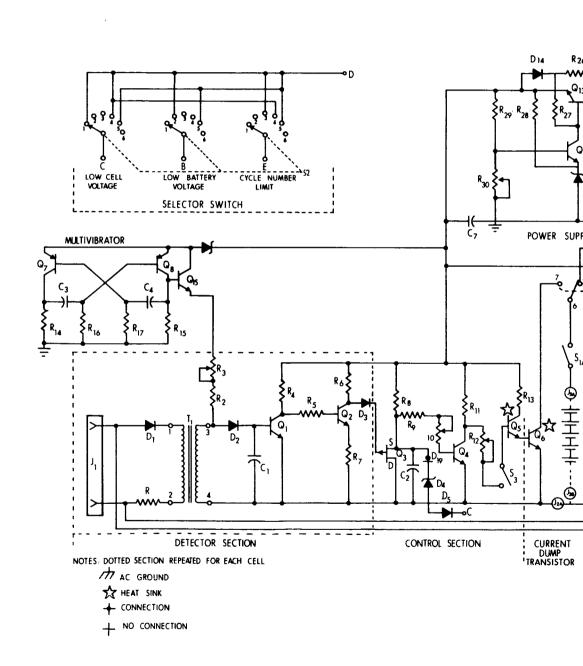
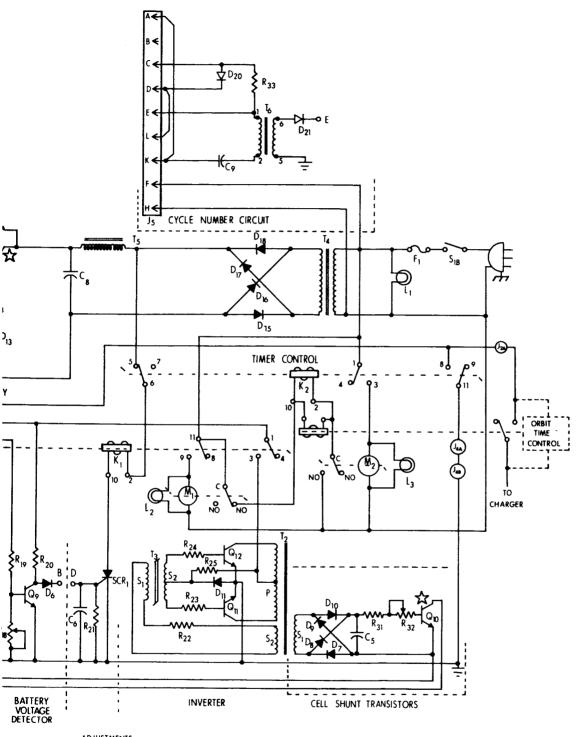


Figure 1-A - Nickel-Cadm

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ADJUSTMENTS

R<sub>10</sub> CUTOFF BAND ADJ R<sub>30</sub> POWER SUPPLY ADJ

 ${\tt R}_{18}$  BATTERY VOLTAGE TRIP ADJ  ${\tt R}_3$  DETECTOR CAL

R12 DUMP CURRENT ADJ

um Battery Reconditioner



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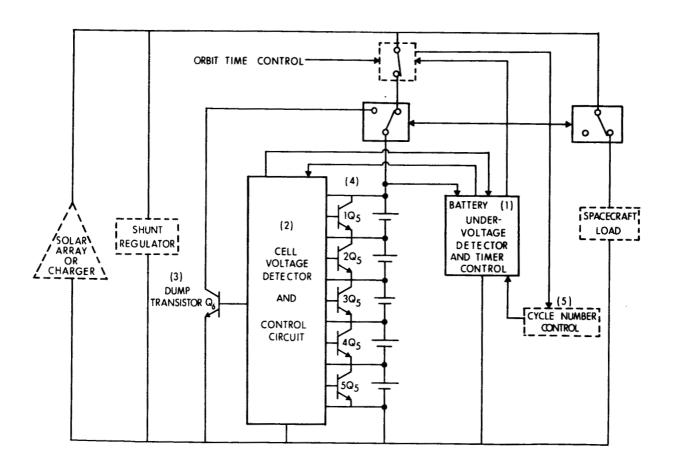


Figure 1
Nickel-Cadmium Battery Reconditioner System

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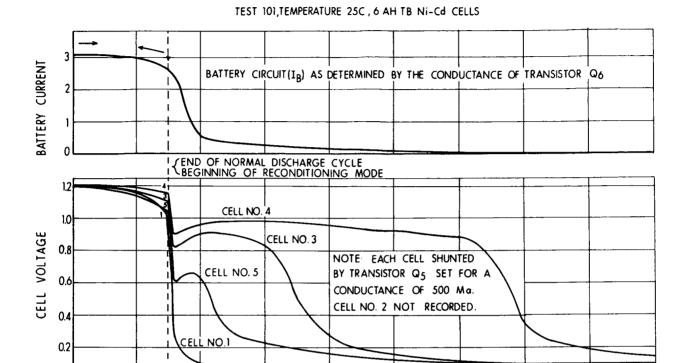


Figure 2
Characteristics of 5-Cell Battery Being
Discharged by Reconditioner

TIME, MINUTES

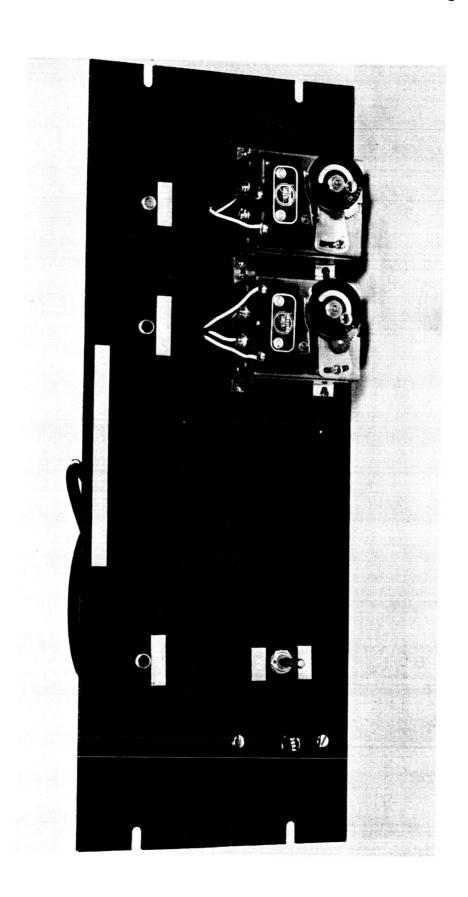


Figure 3 Nickel-Cadmium Battery Reconditioner, Front View

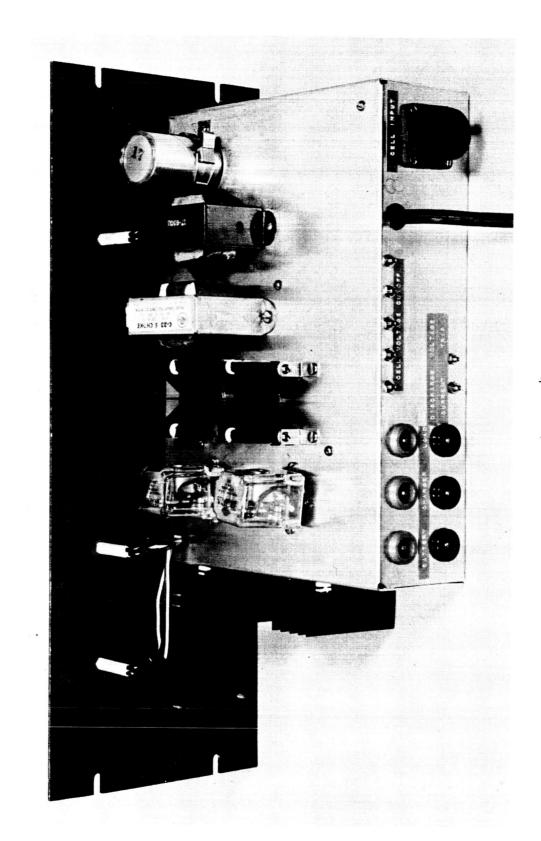


Figure 4 Nickel-Cadmium Battery Reconditioner, Top-Rear View

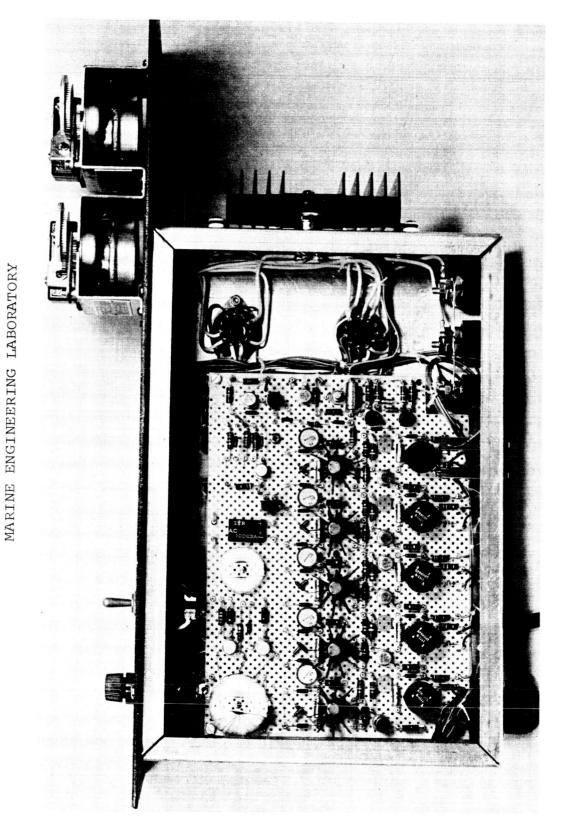


Figure 5 Nickel-Cadmium Battery Reconditioner, Bottom View

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# Appendix A

Operators Guide for Nickel-Cadmium Battery Reconditioner

#### Section I

#### Description

The reference characters used in this description refer to Figure 1-A. The battery terminal under-voltage detector consists of Transistor  $Q_{\rm g}$ , Resistors  $R_{1\,\rm g}$ ,  $R_{1\,\rm g}$ , and  $R_{2\,\rm o}$ . Resistors  $R_{1\,\rm g}$  and  $R_{1\,\rm g}$  form a voltage divider across the battery terminals, with the voltage drop ratio being determined by potentiometer  $R_{1\,\rm g}$ . This potentiometer is adjusted to provide saturation of Transistor  $Q_{\rm g}$  when the battery voltage is above the desired level (approximately 5 volts). If the battery voltage drops below this level, Transistor  $Q_{\rm g}$  comes out of saturation, thus providing an increasing voltage at its collector. Diode  $D_{\rm g}$  becomes forward biased, and allows the collector voltage to appear at Point B. Resistor  $R_{2\,\rm o}$  current limits Transistor  $Q_{\rm g}$  when in saturation and forms a voltage divider with Resistor  $R_{2\,\rm l}$  when Transistor  $Q_{\rm g}$  comes out of saturation.

The Silicon Controlled Rectifier (SCR1) and Relay K1 provide the switching action necessary to place the battery on controlled discharge. Switch Sa provides an "or" connection from the low cell voltage detector, low battery voltage detector, and the cycle number counter to SCR1. The wiring of this switch is such that a low cell, or low terminal voltage, or cycle-number limit, or low cell and cycle-number limit, or low cell and low terminal and cycle-number limit, will determine when the battery is to be placed on controlled discharge. When any one of these conditions exist, a signal appears at Point D, gating SCR, on and energizing the coil of Relay K, thus closing the normally open contacts associated with Relay K1. Three functions are performed by these contacts: (1) The battery is placed across the Discharge Transistor,  $Q_6$ ; (2) The 10 volts d-c supply is connected to the inverter circuit which biases the cell shunt transistors,  $Q_{10}$ , on: (3) The timer motor,  $M_1$ , is energized from the a-c line to start the discharge timer.

With the battery terminals shunted by Transistor  $Q_8$ , the battery discharge current is controlled by the bias condition set on Transistor  $Q_5$  by Resistors  $R_{11}$  and  $R_{13}$ , and Potentioneter  $R_{12}$ . Since Transistor  $Q_4$  is normally cutoff when all cell voltages are above the level set for detector control, the current available through Resistor  $R_{11}$  becomes the base current of Transistors  $Q_5$  and  $Q_6$ , the collector-to-emitter current (battery current) of  $Q_6$  is a direct function of base current of Transistor  $Q_5$ . Potentiometer  $R_{12}$  provides an adjustment for setting

the maximum discharge current from the battery. Switch  $\mathbf{S}_3$  provides a manual control to prevent discharge of the battery by Transistor  $\mathbf{Q}_6$ , when it is desirable to use only the cell shunt transistor for a load.

The individual cell voltage detector consists of Resistors  $R_1 - R_7$ . Diodes  $D_1 - D_3$ , Transistors  $Q_1$  and  $Q_2$ , Capacitor  $C_1$ , and Toroidal Transformer  $T_1$ . This section is repeated for each cell. The square wave signal to drive each detector is provided by a conventional free-running multivibrator. This consists of Resistors  $R_{1.4} - R_{1.7}$ . Transistors  $Q_7$  and  $Q_8$ , and Capacitors  $C_3$  and  $C_4$ . Transistor  $Q_{1.6}$  provides for an impedance transformation between multivibrator and input to detector. The output of the multivibrator is approximately 500 cps, 5-volt square wave, referenced to ground. Diode  $D_{1.2}$  reduces the supply voltage to the multivibrator and may be used to shift the operating range of all the detectors.

Each cell is connected to the input of a detector, Resistor  $R_1$  and Diode  $D_1$ . As a cell voltage decreases, the duc current through Diode D<sub>1</sub> decreases, resulting in an increasing dynamic impedance of the diode. This change in dynamic impedance in the a-c side of Transformer  $T_1$ , which is part of the voltage dividers formed with Resistors R<sub>2</sub> and R<sub>3</sub>. When a cell voltage at the detector input is high, the reflected dynamic impedance in the a.c side of Transformer T, is very small, thereby supporting little voltage drop across the windings. With no current available to the base of Transistor  $Q_1$ , it is in the cutoff region. As the voltage level at Point 3 increases (decreasing cell voltage). Diode Da becomes forward biased and allows energy to be stored in Capacitor C. An increasing voltage level on  $C_1$  causes Transistor  $Q_1$  to come out of cutoff and into the active region. The collector voltage of Transistor  $Q_1$  is then a function of the voltage level at Point 3. voltage at the detector input that determines when Q, comes out of cutoff can be set by Potentiometer Ra.

Since the voltage at the collector of Transistor  $Q_1$  is high when the input to the detector is high (high cell voltage), an inversion is necessary in order to form an "or" connection to the control circuit. Transistor  $Q_2$  with Bias Resistors  $R_6$ ,  $R_6$ , and  $R_7$  provide for this inversion. Diode  $D_3$  from each detector forms the "or" connection to the input of the control circuit. With a high cell voltage at the input to each detector, all detector outputs are near zero.

The output of each detector is connected to the gate of the field effect transistor, Q2, the input to the control circuit. As the output voltage of a detector increases, the voltage at the gate (and the source) of the field effect transistor, Q., increases to Bias Transistor Q, into the active region. The field effect transistor, being almost an ideal voltage control. prevents summing effect from the "or" circuit; i.e. the detector with the highest output voltage will be the controlling detector, irrespective of the lower voltage level of other detectors. Resistor  $R_{\mathbf{e}}$ , and Resistors  $R_{\mathbf{e}}$  and  $R_{\mathbf{i},0}$  are bias resistors for Transistors  $Q_3$  and  $Q_4$ , respectively. With Transistor  $Q_4$  biased into the conducting state and eventually into the saturation region (as cell voltage decreases) Transistors Qs and Qs are gradually cutoff, thus reducing the current from the battery to zero. Resistor R<sub>10</sub> allows the band of actual controlled cutoff to be adjusted for sharp cutoff, or a linear cutoff.

The output of field effect transistor,  $Q_3$  is also connected to the gate of  $SCR_1$  to provide for controlled discharge of the battery, if a cell voltage comes within the active region of the detector before an under-voltage condition occurs on the battery. This prevents a cell from being reversed charged. Field effect diode,  $D_{1\,9}$ , is a current limiting diode to prevent loading the output of Transistor  $Q_3$ , while zener Diode  $D_4$  provides a d-c level shift to compensate for the off-set voltage at the source of the field effect transistor,  $Q_3$ . Diode  $D_5$  is part of the "or" circuit to  $SCR_1$  and prevents interaction from other circuits feeding the "or" connection.

To use the cycle number counter to switch the battery to control discharge, it is necessary to obtain a signal from a set of normally open-normally closed contacts within the counter. Resistor  $R_{33}$ . Transformer  $T_6$ , and Capacitor  $C_9$  form a pulse shaping network that operates from the counter power supply. Upon completing a predetermined number of cycles, the counter resets and a pulse appears at the "or" connection, formed by Diodes  $D_{21}$ ,  $D_{5}$ , and  $D_{6}$ . With Switch  $S_2$  in Position 3, 4, or 5,  $SCR_1$  will be energized, switching the battery to control discharge. Diode  $D_{20}$  is an arc suppressor to protect the contacts of the counter.

The circuit to drive the individual cell load transistors consists of an inverter, a rectifier, and shunt transistor for each cell. The inverter consists of Transistors  $Q_{1,1}$  and  $Q_{1,2}$ , Resistors  $R_{22}-R_{24}$ , Transformers  $T_2$  and  $T_3$ , and Diode  $D_{10}$ . is a conventional two-transformer inverter with the exceptions of Diode D<sub>10</sub> and Resistor R<sub>25</sub> which were added to ensure a positive start when power is applied. The secondary of Transformer T2 consists of five separate windings (one for each cell) so that complete isolation is maintained between cells. bridge rectifier, Diodes  $\mathrm{D_5}$  - $\mathrm{D_8}$ , along with Capacitor  $\mathrm{C_5}$  provides a d-c voltage from the inverter to drive the cell shunt transistor. Resistor R<sub>24</sub> is a current limiting resistor, while Resistor R<sub>25</sub> provides for adjustment of cell discharge currents (50 to 750 ma) by controlling the bias current of the shunt transistor. Each shunt transistor discharges its respective cell to approximately 0 volts. The improtant characteristic of Transistor Q<sub>10</sub> is one having a low saturation voltage. A switching-type transistor serves this application very well.

After sufficient time has been allowed for all cells to be completely discharged, the mechanical actuated switch associated with Timer  $M_1$  closes the normally open contact and applies the a-c line voltage to the coil of Relay  $K_2$ . The contacts associated with Relay  $K_2$  perform three functions: (1) de-energizes the relay coil,  $K_1$  (all contacts return to normally closed position); (2) opens circuit-to-battery load; (3) starts the charge period timer,  $M_2$ , and energizes Relay  $K_3$  to bypass orbit time control. The battery is now charged at a current rate determined by the setting of Timer  $M_2$ . At the end of the charge time, the relay coil of  $K_2$  is de-energized and the battery is returned to normal cycling. Standby status is maintained as long as Switch  $S_1$  is on with the mode of operation being determined by the position of Switch  $S_2$ .

A 115-volt, 60-cps a-c voltage is required to operate this instrument. Timers  $M_1$  and  $M_2$ , and Relays  $K_2$  and  $K_3$  are energized directly from the line of voltage, while Relay  $K_1$  is operated from the output of the bridge rectifier. The d-c power supply consists of a bridge rectifier, Diodes  $D_{15}$ - $D_{18}$ ; a filter, Choke  $T_4$  and Capacitor  $C_8$ ; and the series voltage regulator, Resistors  $R_{26}$ - $R_{30}$ , Diodes  $D_{13}$  and  $D_{14}$ , and Transistors  $Q_{13}$  and  $Q_{14}$ . The 115±10 percent line voltage is converted to 10 volts ±1 percent d-c and is regulated for a 100 percent load charge.

# Section II

# Parts List

Nickel-Cadmium	Battery	Reconditioner

$R_1$	3.3K1	
R <sub>2</sub>	4.7ĸ	
R <sub>3</sub>	25K	1/4-watt, 25-turn potentiometer, wire wound (panel mount)
$R_4$	10K	would (puner mount)
$R_{6}$	510K	
R <sub>e</sub>	10K	
R <sub>7</sub>	2.2K	
R <sub>e</sub>	5.6к	
R <sub>9</sub>	33K	
R <sub>10</sub>	25к	<pre>1/4-watt, 25-turn potentiometer, (circuit board)</pre>
R <sub>11</sub>	lĸ	(322323
R <sub>1 2</sub>	50K	1/4-watt, 25-turn potentiometer, carbon or metal film (panel mount)
R <sub>13</sub>	47Ω	2 watt
R <sub>14</sub> , R <sub>15</sub>	10K	
$R_{16}$ , $R_{17}$	47K	
R <sub>2.8</sub>	2K	1/4-watt, 25-turn potentiometer, carbon or metal film (panel mount)
R <sub>ls</sub> , R <sub>20</sub>	10K	(F /
R <sub>21</sub>	2.7K	

<sup>1</sup> K = thousand

R <sub>2 2</sub>	2.7Ω	
R <sub>23</sub> , R <sub>24</sub>	22Ω	
R <sub>25</sub>	5.6к	
R <sub>26</sub>	560Ω	
R <sub>27</sub>	430Ω	
Ras	620Ω	
R <sub>2 9</sub>	560Ω	
Ra o	2K	1/4-watt, single-turn potentiometer,
R <sub>3 1</sub>	100Ω	carbon, (circuit board)
R <sub>3 2</sub>	2K	<pre>l-watt, single-turn potentiometer (IRC series 100)</pre>
R <sub>3 3</sub>	200K	All resistors 1/2 watt 5 percent unless otherwise noted
		Capacitors
C	10 m f l	10 yr da /Gmmagua Myma 150D)

C <sub>1</sub>	$10 \text{ mf}^1$	10-v dc	(Spraque Type 150D)
C <sup>5</sup>	33 mf	20-v dc	
C3, C4	.033 mf	20-v dc	
C <sub>5</sub>	22 mf	15-v dc	
Ce	10 mf	10-v <b>d</b> c	
$C_{\gamma}$	68 mf	20-v dc	
Ca	1500 mf	25-v dc	
Ce	l mf	200_v đc	

Abbreviations used in this text are from the GPO Style Manual, 1959, unless otherwise noted.

## Transistors

 $Q_1$  2N338

 $Q_{2} = Q_{4} = Q_{14}$  2N1613

Q<sub>3</sub> 2N2609

Q<sub>5</sub>, Q<sub>13</sub> 2N1720

Qs 164-04 (Westinghouse)

 $Q_7 \cdot Q_8$  2N1303

Q<sub>9</sub> 2N1302

 $Q_{10} = Q_{11} = Q_{12}$  2N2192A

Q<sub>15</sub> 2N1308

# Diodes

D<sub>1</sub> lN191

D<sub>2</sub>, D<sub>3</sub>,

 $D_{6} - D_{11}, D_{21}$  1N457

D<sub>4</sub> 1N751A

D<sub>1 2</sub> 2N2192A

D<sub>1 8</sub> ln752

D<sub>1 4</sub> 1N746

 $D_{15} - D_{18}$  1N4002

D<sub>l9</sub> MCL 1300 (Motorla)

D<sub>20</sub> 1N4005

SCR<sub>1</sub> 2N2324

# Transformers and Chokes

T <sub>1</sub>	Spraque Electric Co. Rllll
T <sub>2</sub>	Core, Arnold Engrg. Co. 5504D-1 Primary (2) 47 Turns Bifilar, No. 27 wire Secondary $(S_1)$ (5) 15 Turns, No. 34 wire Secondary $(S_2)$ (1) 15 Turns, No. 34 wire
T <sub>3</sub>	Core, Arnold Engrg. Co. 5515D-1 Primary (1) 50 Turns, No. 34 wire Secondary (2) 35 Turns, Bifilar, No. 34 wire
$\mathbf{T_4}$	Filament Transfer Primary: 117 volts, 50/60 cps Secondary: 26.5 volts, center tapped at 6 amp (Knight No. b-K-48HF)
$T_{5}$	Choke 2 $H^1$ at 200 ma $60\Omega$ d-c resistance (Stanco No. C-2325)
$\mathrm{T}_{\mathbf{g}}$	Pulse Transformer, UTC H-51
	Relays
K <sub>1</sub>	Potter and Brumfield, KRP 14 DG, 24-v d-c Coil
Ka	Potter and Brumfield, KRP 14 AE, 115-v d-c Coil
	Timers
$M_1$ , $M_2$	Industrial Timer Corporation, Model CM 10
	Lights

 $L_1$ ,  $L_2$  Leecraft, No. 36EN2111 RED

Leecraft, No. 36EN2113 Amber

 $L_3$ 

<sup>&</sup>lt;sup>1</sup>H = Henry

## Connectors

J<sub>1</sub> Amphenol, No. MS 3102A-20-33S, with mating plug

 $J_2 - J_4$  Superior Electric Co., No. RP25GR (3)

J<sub>5</sub> Amphenol, No. MS 3102A-28-14P, with mating socket

## Switch

S<sub>1</sub> DPST - Toggle, 10-amp contacts

S<sub>2</sub> 3 pole - 5-position; nonshort

S<sub>3</sub> SPST Toggle

## Heat Sink

Wakefield No. NF 209 (5)

Wakefield No. NC 403K (1)

Wakefield No. NC 30lm (2)

## Power Cord

Belden N 17408-S, 3 wire, 8 ft

#### Section III

## Adjustment and Calibration Procedure

Note: It is recommended that all adjustment procedures be read before proceeding with individual adjustments.

The following instruments are necessary for adjustment and calibration:

- 1 Multirange voltmeter (digital voltameter preferred)
- 2 Power supply; 0-5 volts, 0-750 milliamps
- 3 Power supply; 0-10 volts, 0-5 amperes
- 4 Ammeter; multirange clip-on, 5 amperes maximum
- A. Power Supply Voltage Adjustment
  - 1 Plug instrument in ac (115 volts, 60 cps) and place Switch  $S_1$  in "on" position.
  - 2 With a voltmeter across Capacitor  $C_7$ , adjust Resistor  $R_{3,0}$  (on circuit board) for 10 volts on meter.
- B. Battery Low-Voltage Trip-Level Adjustment.

Note: Complete A above before preceeding.

- l Set cams on Timer  $M_1$  and  $M_2$  at beginning of lobe (cams rotate clockwise).
- 2 Set Selector Switch S2 to "low battery."
- 3 Connect power supply to battery terminals and set voltage to desired trip level.
- 4 Adjust battery voltage trip adjustment screw  $(R_{18})$  until Relay  $K_1$  is energized (discharge light comes on).
- 5 Reset of Relay  $K_1$  may be accomplished by moving the cam of Timer  $M_1$  to it's flank.

C. Battery Dump Transistor Current Adjustment.

Note: Complete A and B above before preceeding.

- 1 Place cell detector  $(J_1)$  inputs in parallel and connect to approximately 2-volt power supply.
- 2 Place dump transistor switch  $(S_3)$  in "on" position.
- 3 Adjust dump current adjustment screw (R<sub>12</sub>) for desired current. A clip-on ammeter may be used to measure this current.
- D. Minimum Cell Voltage for Zero Dump Current Adjustment.

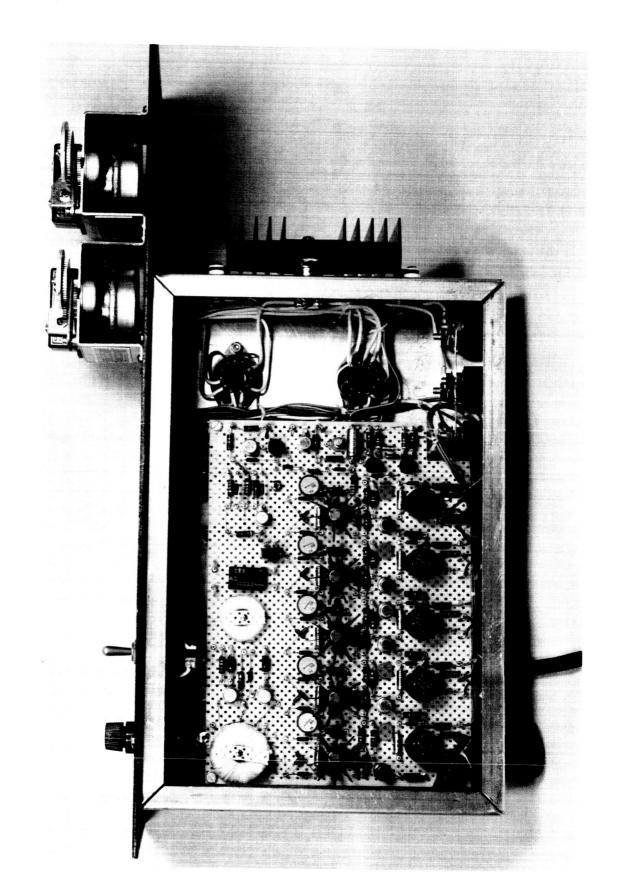
Note: This procedure may be a continuation of C above.

- 1 Connect one detector input to a power-supply voltage set at the level desired for zero dump current. (The other detectors are in parallel across the 2-volt supply.)
- 2 Set cutoff band adjustment screw  $(R_{10})$  at midrange of adjustment. This may be varied to obtain desired cutoff band in conjunction with the following steps:
  - (a) Adjust the detector calibration screw, corresponding to the detector connected to the variable power supply for zero dump current using a sensitive ammeter (clip-on type) to indicate zero current.
  - (b) Vary the input to the detector and adjust cutoff band adjustment until desired "active" band of detector control over transistor dump current is obtained.
  - (c) Repeat Steps (a) and (b) above for each detector until all 5 detectors are calibrated.

E. Individual Cell Discharge Adjustment.

Note: This adjustment may be made in conjunction with D-l above.

- 1 Set the voltage input to the single detector to 1 volt.
- 2 Connect a clip-on ammeter on a lead to the detector input for current measurement.
- 3 Adjust Potentiometer  $R_{32}$  (on circuit board) for corresponding detector to obtain the desired current level to be shunted by the individual cell shunt transistor.
- 4 Repeat Steps (1), (2), and (3) above for all five detectors or shunt transistors.



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A device has been designed for automatically reconditioning a nickel-cadmium battery. Individual cell voltage detectors and/or a battery terminal voltage are utilized to determine when reconditioning is necessary. Reconditioning is accomplished by discharging each cell to nearly 0 volts, then recharging the battery at a small current. Controlled discharge of the battery is accomplished by individual cell shunting transistors and a battery current shunt transistor. Cell reversal is prevented by cutting off the battery dump current when a cell voltage reaches a low limit.

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